ICTP-ITU-URSI School on Wireless Networking for Development
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Basic Antenna Theory

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Note: These are preliminary notes, intended only for distribution among the participants. Beware of misprints!

Purpose

- to refresh basic concepts related to the antenna physics
 - needed to understand better the operation and design of microwave links and networks

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Outline

- Introduction
- Review of basic antenna types
- Radiation pattern, gain, polarization
- Equivalent circuit & radiation efficiency
- Smart antennas
- Some theory
- Summary

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Quiz

Transmitting antennas are used to radiate energy in the form of radio waves

Receiving antennas -- to capture that energy

Somebody told that the receiving antenna during the reception also radiates radio waves

Is it a true fact or a slip of the tongue?

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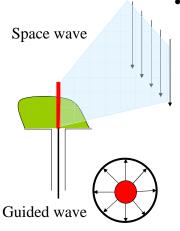
Intended & unintended radiators

- Intended antennas
 - To produce/ receive specified EM waves: Radiocommunication antennas; Measuring antennas; EM sensors, probes; EM applicators (Industrial, Medical, Scientific)
- · Unintended antennas active
 - EM waves radiated as an unintended side-effect:
 - Any conductor/ installation with varying electrical current (e.g. electrical installation of vehicles)
 - · Any slot/ opening in the screen of a device/ cable carrying RF current
 - Any discontinuity in transmission medium (e.g. conducting structures/ installations) irradiated by EM waves
- Unintended antennas passive
 - Stationary (e.g. antenna masts or power line wires); Time-varying (e.g. windmill or helicopter propellers); Transient (e.g. aeroplanes, missiles)

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Antenna fuction



- Transformation of a guided EM wave (in waveguide/ transmission line) into an EM wave freely propagating in space, with specified directional characteristics (or vice versa)
 - Transformation from time-function in one-dimensional space into timefunction in three dimensional space
 - The specific form of the radiated wave is defined by the antenna structure and the environment

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Transmission line

 Power transport medium - must avoid power reflections, otherwise use matching devices

Radiator

 Must radiate efficiently – must be of a size comparable with the half-wavelength

Resonator

Kraus p.2

Unavoidable - for broadband applications resonances must be attenuated

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Monopole (dipole over plane) Sharp Thin radiator transition Thick radiator region Smooth transition region High-Q Uniform wave Narrowband traveling Low-Q along the line Broadband If there is an inhomogeneity (obstacle, or sharp transition), higher field-modes, reflections, and standing wave appear. With standing wave, the energy is stored in, and oscillates from electric energy to magnetic one and back. This can be modeled as a resonating LC circuit with $Q = (energy \ stored \ per \ cycle) / (energy \ lost \ per \ cycle)$

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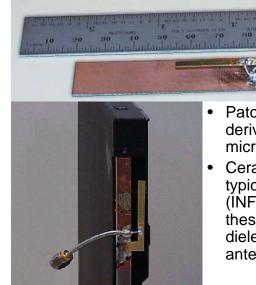
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Antennas for laptop applications | Monopole | Dipole | D



Source: D. Liu et al.: Developing integrated antenna subsystems for laptop computers: IBM J. RES. & DEV. VOI. 47 NO. 2/3 MARCH/MAY 2003 p. 355-367

 Patch and slot antennas derived from printed-circuit and micro-strip technologies

 Ceramic chip antennas are typically helical or inverted-F (INF) antennas, or variations of these two types with high dielectric loading to reduce the antenna size

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Slot & INF antennas

- Slot antenna: a slot is cut from a large (relative to the slot length) metal plate.
 - The center conductor of the feeding coaxial cable is connected to one side of the slot, and the outside conductor of the cable - to the other side of the slot.
- The slot length is some (λ /2) for the slot antenna and (λ /4) long for the INF antenna.
- The slot and INF antennas behave similarly.
 - The slot antenna can be considered as a loaded version of the INF antenna. The load is a quarter-wavelength stub, i.e. a narrowband device.
 - When the feed point is moved to the short-circuited end of the slot (or INF) antenna, the impedance decreases. When it is moved to the slot center (or open end of the INF antenna), the impedance increases

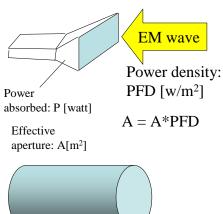
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Example double-layer printed Yagi antenna driver dielectric substrate transformer top layer reflector bottom layer top layer microstrip Note: no galvanic contact with the director reflector bottom layer Source: N Gregorieva Property of Trouwant 13

- Patch and slot antennas are
 - Cheap and easy to fabricate and to mount
 - Suited for integration
 - Light and mechanically robust
 - Have low cross-polarization
 - Low-profile widely used in antenna arrays
 - spacecrafts, satellites, missiles, cars and other mobile applications

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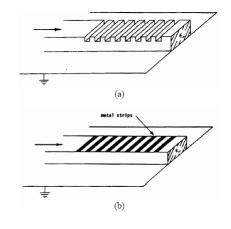


- Aperture antennas derived from waveguide technology (circular, rectangular)
- Can transfer high power (magnetrons, klystrons)
- · Above few GHz
- Will be explored in practice during the school
- Note: The aperture concept is applicable also to wired antennas. For instance, the max effective aperture of linear $\lambda/2$ wavelength dipole antenna is $\lambda^2/8$

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Leaky-wave antennas



- Derived from millimeterwave guides (dielectric guides, microstrip lines, coplanar and slot lines).
- For frequencies > 30 GHz, including infrared
- Subject of intensive study.
 - Note: Periodical discontinuities near the end of the guide lead to substantial radiation leakage (radiation from the dielectric surface).

Source: adapted from N Gregorieva

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Reflector antennas

- Reflectors are used to concentrate flux of EM energy radiated/ received, or to change its direction
- Usually, they are parabolic (paraboloidal).
 - The first parabolic (cylinder) reflector antenna was used by Heinrich Hertz in 1888.
- · Large reflectors have high gain and directivity
 - Are not easy to fabricate
 - Are not mechanically robust
 - Typical applications: radio telescopes, satellite telecommunications.

Source: adapted from N Gregorieva

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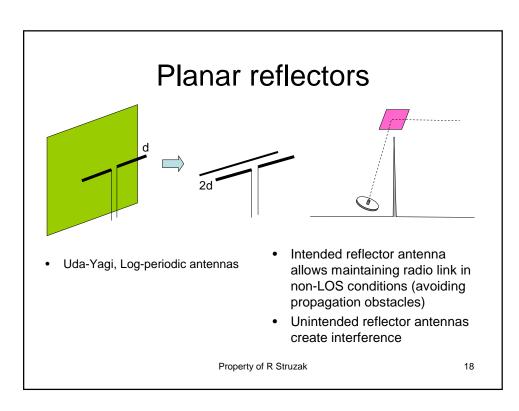
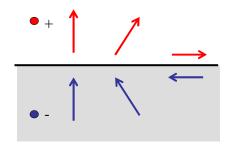


Image Theory

- Antenna above perfectly conducting plane surface
- Tangential electrical field component = 0
 - vertical components: the same direction
 - horizontal components: opposite directions
- The field (above the ground) is the same as if the ground is replaced by an mirror image of the antenna
- http://www.amanogawa.com/archive/wavesA.html

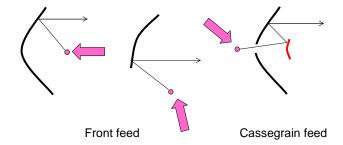


Elliptical polarization: change of the rotation sense!

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Paraboloidal reflectors



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The largest radio telescopes

- Max Plank Institüt für Radioastronomie radio telescope, Effelsberg (Germany), 100-m paraboloidal reflector
- The Green Bank Telescope (the National Radio Astronomy Observatory) – paraboloid of aperture 100 m

Source: adapted from N Gregorieva

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The Arecibo Observatory Antenna System



The world's largest single radio telescope

304.8-m spherical reflector National Astronomy and lonosphere Center (USA), Arecibo, Puerto Rico

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The Arecibo Radio Telescope

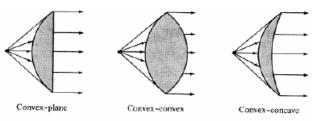


[Sky & Telescope Feb 1997 p. 29]

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Lens antennas



(a) Lens antennas with index of refraction n > 1

Lenses play a similar role to that of reflectors in reflector antennas: they collimate divergent energy

Often preferred to reflectors at frequencies > 100 GHz.

Source: Kraus p.382, N Gregorieva

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- Antenna characteristics of gain, beamwidth, efficiency, polarization, and impedance are independent of the antenna's use for either transmitting or receiving.
- The properties we will discuss here apply to both cases.

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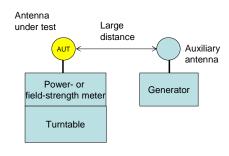
Radiation pattern

- The radiation pattern of antenna is a representation (pictorial or mathematical) of the distribution of the power out-flowing (radiated) from the antenna (in the case of transmitting antenna), or inflowing (received) to the antenna (in the case of receiving antenna) as a function of direction angles from the antenna
 - Antenna radiation pattern (antenna pattern):
 - is defined for large distances from the antenna, where the spatial (angular) distribution of the radiated power does not depend on the distance from the radiation source
 - is independent on the power flow direction: it is the same when the antenna is used to transmit and when it is used to receive radio waves
 - is usually different for different frequencies and different polarizations of radio wave radiated/ received

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Power pattern vs. Field pattern



- The power pattern and the field patterns are inter-related for plane wave:
 P(θ, φ) = (1/η)*|E(θ, φ)|² = η*|H(θ, φ)|²
 - P = power
 - E = electrical field component vector
 - H = magnetic field component vector
 - $\eta = 377 \text{ ohm (free-space, plane wave} \\ \text{impedance)} \\ \text{Property of R Struzak}$

 The power pattern is the measured (calculated) and plotted received power: |P(θ, φ)| at a constant (large) distance from the antenna

The amplitude field pattern is the measured (calculated) and plotted electric (magnetic) field intensity, |E(θ, φ)| or |H(θ, φ)| at a constant (large) distance from the antenna

Normalized pattern

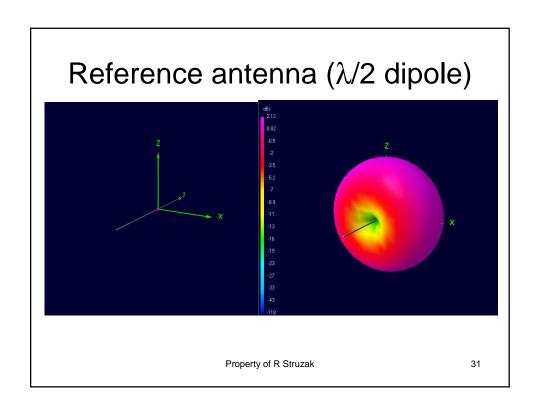
- Usually, the pattern describes the normalized field (power) values with respect to the maximum value.
 - Note: The power pattern and the amplitude field pattern are the same when computed and when plotted in dB.

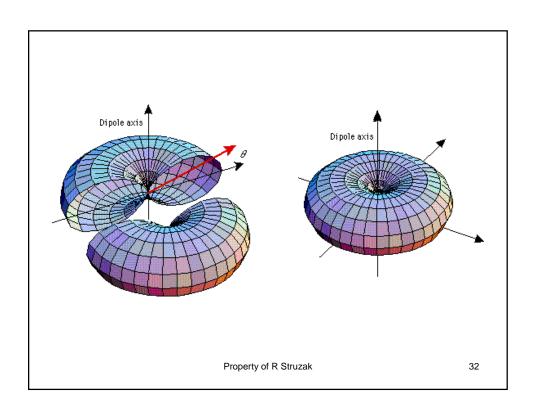
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Reference antenna ($\lambda/2$ dipole)

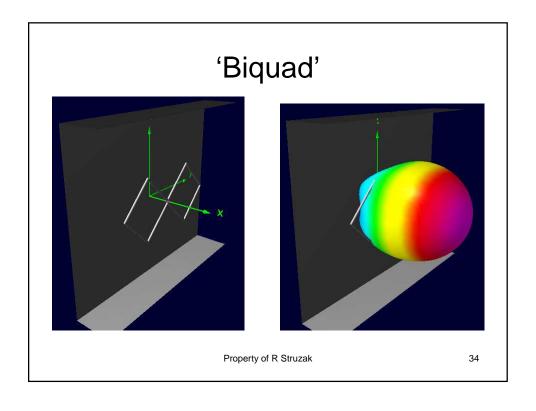
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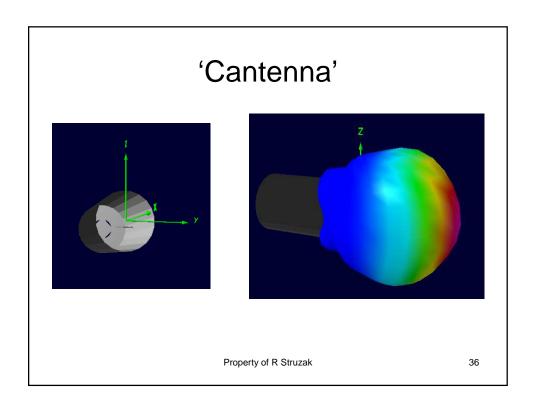
Biquad antenna

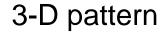
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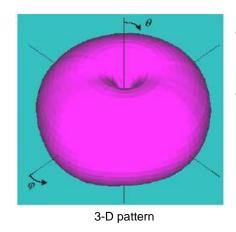


Cantenna

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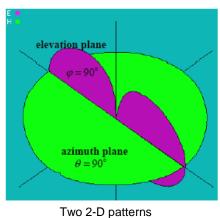
- Antenna radiation pattern is 3-dimensional
- The 3-D plot of antenna pattern assumes both angles θ and φ varying, which is difficult to produce and to interpret

Source: NK Nikolova

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2-D pattern



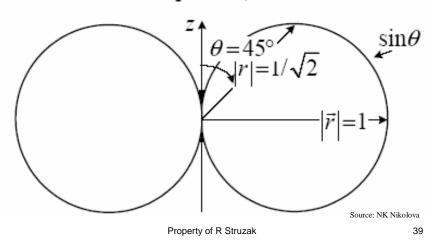
- Usually the antenna pattern is presented as a 2-D plot, with only one of the direction angles, θ or φ varies
- It is an intersection of the 3-D one with a given plane
 - usually it is a θ = const plane or a φ= const plane that contains the pattern's maximum

Source: NK Nikolova

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Example: a short dipole on z-axis

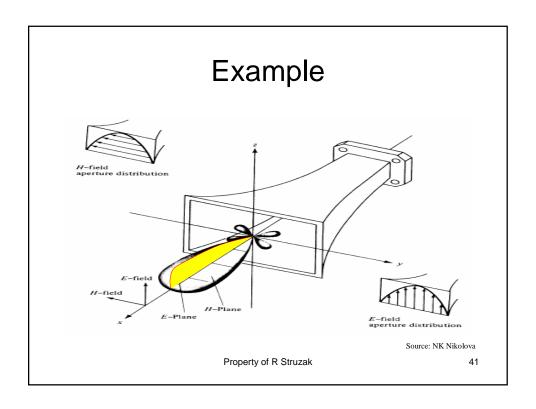
Elevation plane: $\varphi = const$



Principal patterns

- Principal patterns are the 2-D patterns of linearly polarized antennas, measured in 2 planes
 - the *E-plane:* a plane parallel to the *E* vector and containing the direction of maximum radiation, and
 - 2. the *H-plane:* a plane parallel to the *H* vector, orthogonal to the *E*-plane, and containing the direction of maximum radiation

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Isotropic antenna



- Isotropic antenna or isotropic radiator is a hypothetical (not physically realizable) concept, used as a useful reference to describe real antennas.
- Isotropic antenna radiates equally in all directions.
 - Its radiation pattern is represented by a sphere whose center coincides with the location of the isotropic radiator.

Source: NK Nikolova

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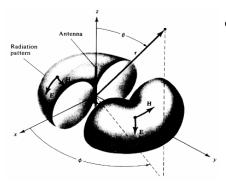
Directional antenna

- Directional antenna is an antenna, which radiates (or receives) much more power in (or from) some directions than in (or from) others.
 - Note: Usually, this term is applied to antennas whose directivity is much higher than that of a half-wavelength dipole.

Source: NK Nikolova

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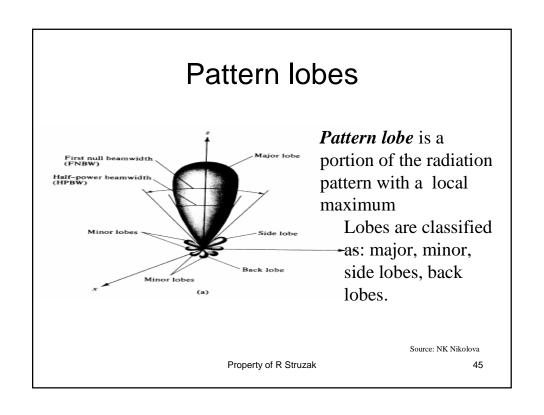
Omnidirectional antenna

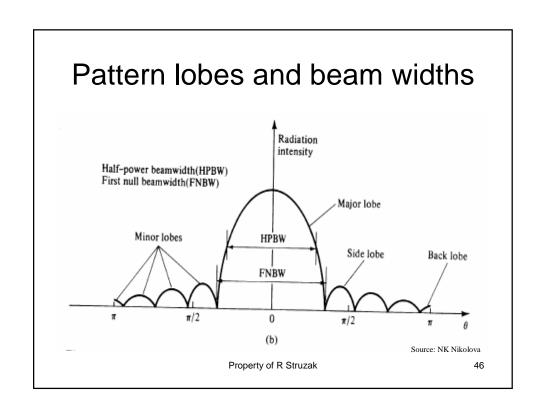


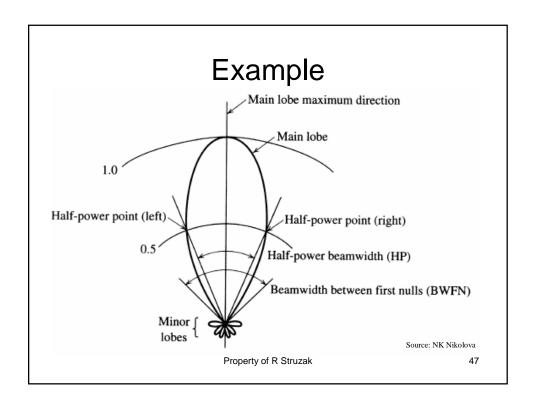
- An antenna, which has a nondirectional pattern in a plane
 - It is usually directional in other planes

Source: NK Nikolova

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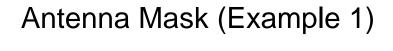


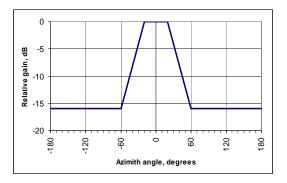
Beamwidth

- Half-power beamwidth (HPBW) is the angle between two vectors from the pattern's origin to the points of the major lobe where the radiation intensity is half its maximum
 - Often used to describe the antenna resolution properties
 » Important in radar technology, radioastronomy, etc.
- *First-null beamwidth* (FNBW) is the angle between two vectors, originating at the pattern's origin and tangent to the main beam at its base.

» Often FNBW ≈ 2*HPBW

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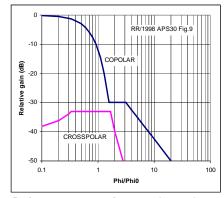
Typical relative directivity- mask of receiving antenna (Yagi ant., TV dcm waves)

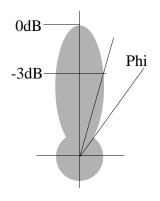
[CCIR doc. 11/645, 17-Oct 1989)

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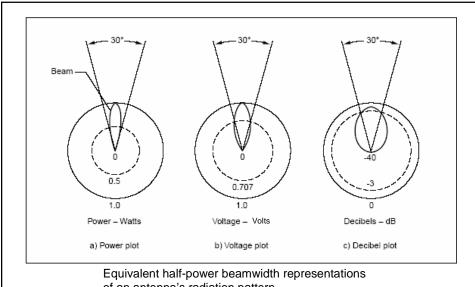
Antenna Mask (Example 2)





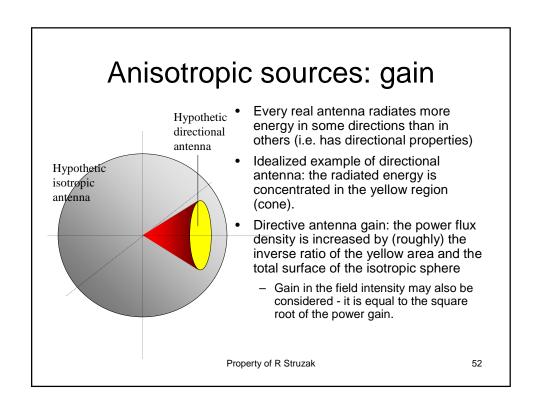
Reference pattern for co-polar and cross-polar components for satellite transmitting antennas in Regions 1 and 3 (Broadcasting ~12 GHz)

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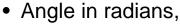


of an antenna's radiation pattern.

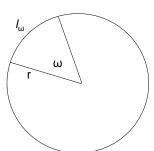
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Plane angle: radian



 $\omega = I_{\omega} / r;$ $I_{\omega} = \omega^* r$



- $-I_{\omega}$ is the length of the arc segment supported by the angle ω in a circle of radius r.
- There are 2π rad in a full circle
- $-1 \text{ rad} = (360 / 2\pi) \text{ deg}$

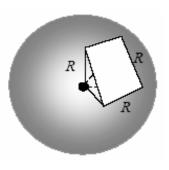
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Solid angle: steradian

• Solid angle in steradians (sr),

$$\Omega = (S_O)/r^2;$$
 $S_O = \Omega r^2$

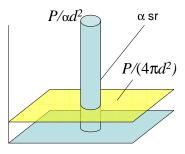


 ${\bf S}_\Omega$ is the spherical surface area supported by the solid angle Ω in a sphere of radius r

- The steradian is the area cut out by the solid angle, divided by the sphere's radius squared - 'squared radian'.
- If the area is S, and the radius is d, then the angle is S/d² steradians. The total solid angle (a full sphere) is thus 4π steradians.
- As one radian is $180/\pi = 57.3$ degrees, the total solid angle is $4\pi \times (57.3)^2 \approx 41253$ square degrees, one steradian is 3282.806 square degrees, and one square degree is about 305 x 10-6 steradians

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Example: gain of 1 deg² antenna



 $G = 4\pi/\alpha$

If $\alpha = 1 \text{ deg2}$, then

 $G = 4\pi/305*10-6 = 46 \text{ dB}$

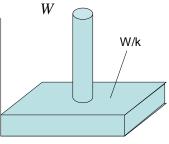
- A hypothetical source radiates P watts uniformly within the solid angle of α steradians in a given direction and zero outside
- The total surface of the sphere is 4πd² and the average irradiance is the power divided by the surface: [P/(4πd²)] w/m²
- α steradians corresponds to spherical surface of αc^P and irradiance within that angle is [P/αc^P] w/m²
- The antenna gain equals the ratio of these two, or $4\pi/\alpha$
- For $\alpha = 1 \text{ deg}^2$ (= 305*10-6 sr); the gain = $4\pi/305*10$ -6 = 46 dB. ,

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Effect of sidelobes

Let the main beamwidth of an antenna be Ω square degrees, with uniform irradiance of W watts per square meter. Let the sidelobe irradiance (outside the main beam) be uniform and k times weaker, i.e. (W/k) watts per square meter, k \geq 1. Then:



$$G = \frac{W}{W_0} = \frac{1}{\frac{1}{k} + \left(\frac{k-1}{k}\right)\left(\frac{\Omega}{41253}\right)}$$

The gain decreases with the sidelobe level and beamwidth.

If the main lobe is 1 square degree and the sidelobes are attenuated by 20 dB, then k = 100 and G = 100 (or 20dB), much less than in the previous example (46dB).

In the limit, when k = 1, the gain tends to 1 and antenna becomes isotropic.

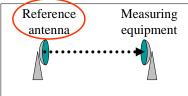
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$$\begin{split} P_M &= W\Omega d^2 & \text{- power radiated within the main lobe} \\ P_S &= \left(\frac{W}{k}\right) (41253 - \Omega) d^2 & \text{- power radiated by sidelobes} \\ P_T &= P_M + P_S = W d^2 \left(\Omega + \frac{41253}{k} - \frac{\Omega}{k}\right) & \text{- total power} \\ W_0 &= \frac{P_T}{41253 d^2} = W \left[\frac{1}{k} + \frac{k\Omega - \Omega}{41253k}\right] & \text{- average irradiation} \\ G &= \frac{W}{W_0} = \frac{1}{\frac{1}{k} + \left(\frac{k-1}{k}\right) \left(\frac{\Omega}{41253}\right)} & \text{- antenna gain} \end{split}$$

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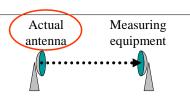
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Antenna gain measurement



 $\begin{array}{ll} P_o = \text{Power} & S_0 = \text{Power} \\ \text{delivered to} & \text{received} \\ \text{the reference} & \text{(the same in} \\ \text{antenna} & \text{both steps)} \end{array}$

Step 1: reference



 $\begin{array}{ll} P = Power \\ \text{delivered to} \\ \text{the actual} \\ \text{antenna} \end{array} \hspace{0.5cm} \begin{array}{ll} S = Power \\ \text{received} \\ \text{(the same in both steps)} \end{array}$

Step 2: substitution

Antenna Gain = $(P/P_0)_{S=S0}$

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Antenna Gains G_i, G_d

- Unless otherwise specified, the gain refers to the direction of maximum radiation.
- Gain is a dimension-less factor related to power and usually expressed in decibels
- G_i "Isotropic Power Gain" theoretical concept, the reference antenna is isotropic
- G_d the reference antenna is a half-wave dipole

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Typical Gain and Beamwidth

Type of antenna	G _i [dB]	BeamW.
Isotropic	0	360°x360°
Half-wave Dipole	2	360°x120°
Helix (10 turn)	14	35°x35°
Small dish	16	30°x30°
Large dish	45	1ºx1º

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Gain, Directivity, Radiation Efficiency

- The radiation intensity, directivity and gain are measures of the ability of an antenna to concentrate power in a particular direction.
- $G(\vartheta, \varphi) = \eta D(\vartheta, \varphi)$ $\eta = \frac{P_T}{P_0}$

(0.5 - 0.75)

η: radiation efficiency

- Directivity relates to the power radiated by antenna
- Gain relates to the power delivered to antenna (P_T) Property of R Struzak

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Antenna gain and effective area

- Measure of the effective absorption area presented by an antenna to an incident plane wave.
- Depends on the antenna gain and wavelength

$$A_e = \eta \frac{\lambda^2}{4\pi} G(\theta, \varphi) \text{ [m}^2\text{]}$$

Aperture efficiency: $\eta_a = A_e \ / \ A$ A: physical area of antenna's aperture, square meters

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Power Transfer in Free Space

$$\begin{split} P_R &= PFD \cdot A_e \\ &= \left(\frac{G_T P_T}{4\pi r^2}\right) \!\! \left(\frac{\lambda^2 G_R}{4\pi}\right) \\ &= P_T G_T G_R \!\! \left(\frac{\lambda}{4\pi r}\right)^2 \end{split} \quad \begin{array}{l} \bullet \quad P_R \text{: power available at the receiving antenna} \\ \bullet \quad P_T \text{: power delivered to the transmitting antenna} \\ \bullet \quad G_R \text{: gain of the transmitting antenna} \\ \bullet \quad G_T \text{: gain of the receiving antenna} \\ \bullet \quad G_T \text{:$$

- wavelength [m]
- P_R: power available at the receiving antenna
- P_T: power delivered to the
- antenna in the direction of the receiving antenna
- G_T: gain of the receiving antenna in the direction of the transmitting antenna
- Matched polarizations

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e.i.r.p.

 Equivalent Isotropically Radiated Power (in a given direction):

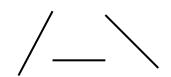
$$e.i.r.p. = PG_i$$

• The product of the power supplied to the antenna and the antenna gain (relative to an isotropic antenna) in a given direction

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Linear Polarization

 In a linearly polarized plane wave the direction of the E (or H) vector is constant.



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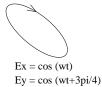
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Elliptical Polarization

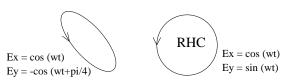




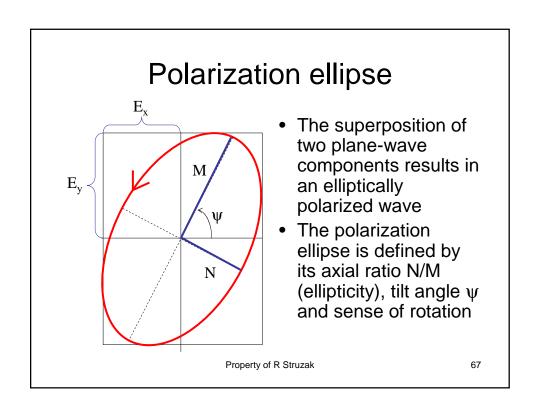


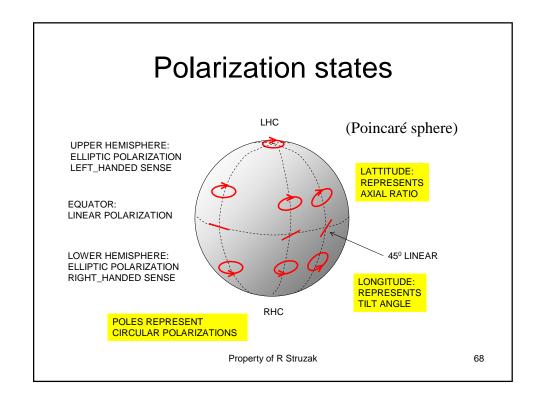


Ex = cos (wt)Ey = -sin (wt)



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Comments on Polarization

- At any moment in a chosen reference point in space, there is actually a single electric vector E (and associated magnetic vector H).
- This is the result of superposition (addition) of the instantaneous fields E (and H) produced by all radiation sources active at the moment.
- The separation of fields by their wavelength, polarization, or direction is the result of 'filtration'.

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Antenna Polarization

 The polarization of an antenna in a specific direction is defined to be the polarization of the wave produced by the antenna at a great distance at this direction

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Polarization Efficiency

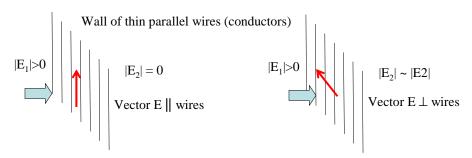
- The power received by an antenna from a particular direction is maximal if the polarization of the incident wave and the polarization of the antenna in the wave arrival direction have:
 - the same axial ratio
 - the same sense of polarization
 - the same spatial orientation

.

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Polarization filters/ reflectors



Wire distance $\sim 0.1\lambda$

 At the surface of ideal conductor the tangential electrical field component = 0

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Outline

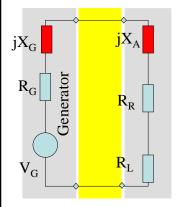
- Introduction
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Transmitting antenna equivalent circuit Antenna Radio wave Transm. line Transmitter The transmitter with the transmission line is represented by an (Thevenin) equivalent generator jX_A jX_G The antenna is represented by its input impedance (which is frequency-dependent and is influenced by objects nearby) as seem from the generator jX_A represents energy stored in electric (E_e) and magnetic (E_m) near-field components; if $|Ee| = |E_m|$ then $X_A = 0$ (antenna resonance) R_r R_r represents energy radiated into space (far-field components) R₁ represents energy lost, i.e. transformed into heat in the antenna structure Property of R Struzak 74

Power transfer: Tx antenna



Transmitter is represented by an eqivalent generator with V_G , R_G , $X_G = const$.

Let
$$R_A = R_R + R_L$$
; $R_A, X_A = \text{var}$.

The power absorbed by antenna $P = I^2 R_A$

$$I^{2} = \left[\frac{V_{G}}{\sqrt{(R_{G} + R_{A})^{2} + (X_{G} + X_{A})^{2}}} \right]^{2}$$

$$P = V_G^2 \frac{R_A}{(R_G + R_A)^2 + (X_G + X_A)^2}$$

$$P = V_G^2 \frac{R_A}{\left(R_G + R_A\right)^2 + \left(X_G + X_A\right)^2}$$

$$P = \left(\frac{V_G^2}{R_G}\right) \frac{\frac{R_A}{R_G}}{\left(1 + \frac{R_A}{R_G}\right)^2 + \left(\frac{X_G}{R_G} + \frac{X_A}{R_G}\right)^2}$$

$$P = V_G^2 \frac{R_A}{(R_G + R_A)^2 + X_G^2 + 2X_G X_A + X_A^2}$$

$$\frac{\partial P}{\partial X_A} = V_G^2 \left(-\frac{R_A (2X_G + 2X_A)}{\left[(R_G + R_A)^2 + (X_G + X_A)^2 \right]^2} \right)$$

$$\frac{\partial P}{\partial X_A} = 0, \text{ when } X_A = -X_G$$

$$Let X_G + X_A = 0. \text{ Then } P = V_G^2 \frac{R_A}{(R_G + R_A)^2} - R_A 2 (R_G + R_A)^2 \right]$$

$$\frac{\partial P}{\partial R_A} = V_G^2 \left(\frac{(R_G + R_A)^2 - R_A 2 (R_G + R_A)}{\left[(R_G + R_A)^2 \right]^2} \right) = V_G^2 \left(\frac{R_G^2 + 2R_G R_A + \overline{R_A^2} - 2R_G R_A - 2\overline{R_A^2}}{\left[(R_G + R_A)^2 - \overline{R_A^2} - \overline{R_A^2} - \overline{R_A^2} \right]} \right)$$

$$\begin{aligned} & \textit{Maximum} : \frac{\partial P}{\partial R_A} + \frac{\partial P}{\partial X_A} = 0 \\ & R_A = R_G, \quad X_A = -X_G \\ & P = \frac{V_G^2}{4R_G} \end{aligned}$$

$$P = V_G^2 \frac{R_A}{\left(R_G + R_A\right)^2 + X_G^2 + 2X_G X_A + X_A^2}$$

$$\frac{\partial P}{\partial X_A} = V_G^2 \left(-\frac{R_A \left(2X_G + 2X_A\right)}{\left[\left(R_G + R_A\right)^2 + \left(X_G + X_A\right)^2\right]^2} \right)$$

$$\frac{\partial P}{\partial X_A} = 0, \text{ when } X_A = -X_G$$

$$\frac{\partial P}{\partial X_A} = 0, \text{ when } X_A = -X_G$$

$$= V_G^2 \left(\frac{R_G + R_A^2 - R_A^2 \left(R_G + R_A^2\right)}{\left[\left(R_G + R_A^2\right)^2 - R_A^2 \left(R_G + R_A^2\right)^2\right]^2} \right)$$

$$= V_G^2 \left(\frac{R_G^2 + \overline{2R_G R_A} + \overline{R_A^2} - \overline{2R_G R_A} - 2\overline{R_A^2}}{\left[\left(R_G + R_A^2\right)^2\right]^2} \right)$$

$$\frac{\partial P}{\partial R_A} = 0, \text{ when } R_G = R_A$$

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Impedance matching

$$R_{A} = R_{r} + R_{l} = R_{g}$$

$$X_{A} = -X_{g}$$

$$P_{A} = \frac{\left|V_{g}\right|^{2}}{4R_{A}}$$

$$P_{g} = \frac{\left|V_{g}\right|^{2}}{4R_{g}} \quad (=P_{A})$$

$$P_{r} = P_{A} \frac{R_{r}}{\left(R_{r} + R_{l}\right)}$$

$$P_{l} = P_{A} \frac{R_{l}}{\left(R_{r} + R_{l}\right)}$$

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Power vs. field strength

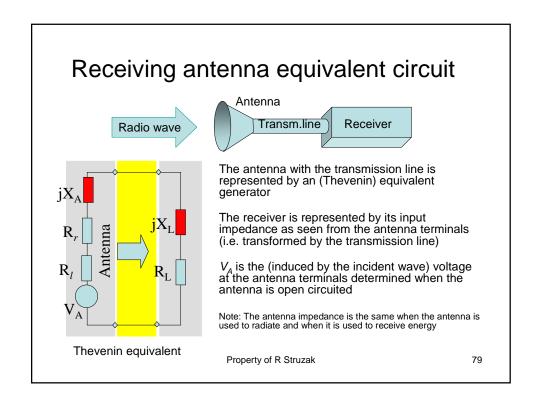
$$P_r = \frac{\left|E\right|^2}{Z_0} \rightarrow \left|E\right| = \sqrt{P_r Z_0}$$

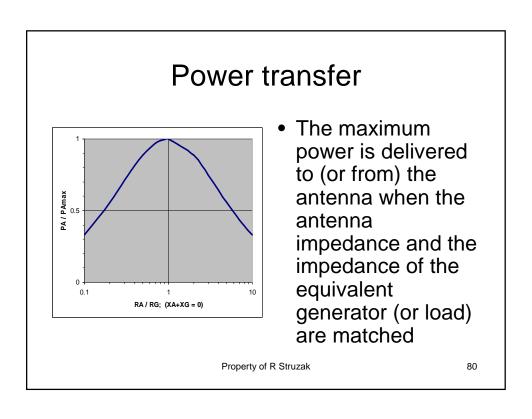
$$\left|E\right| = \sqrt{E_\theta^2 + E_\phi^2}$$

$$\left|H\right| = \frac{\left|E\right|}{Z_0}$$

$$Z_0 = 377 \text{ ohms}$$
for plane wave
in free space

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- · When the impedances are matched
 - Half of the source power is delivered to the load and half is dissipated within the (equivalent) generator as heat
 - In the case of receiving antenna, a part (P_i) of the power captured is lost as heat in the antenna elements, , the other part being reradiated (scattered) back into space
 - Even when the antenna losses tend to zero, still only half of the power captured is delivered to the load (in the case of conjugate matching), the other half being scattered back into space

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- When the antenna impedance is not matched to the transmitter output impedance (or to the receiver input impedance) or to the transmission line between them, impedance-matching devices must be used for maximum power transfer
- Inexpensive impedance-matching devices are usually narrow-band
- Transmission lines often have significant losses

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Radiation efficiency

- The radiation efficiency e indicates how efficiently the antenna uses the RF power
- It is the ratio of the power radiated by the antenna and the total power delivered to the antenna terminals (in transmitting mode). In terms of equivalent circuit parameters:

$$e = \frac{R_r}{R_r + R_l}$$

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Outline

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Antenna arrays

- Consist of multiple (usually identical) antennas (elements) 'collaborating' to synthesize radiation characteristics not available with a single antenna. They are able
 - to match the radiation pattern to the desired coverage area
 - to change the radiation pattern electronically (electronic scanning) through the control of the phase and the amplitude of the signal fed to each element
 - to adapt to changing signal conditions
 - to increase transmission capacity by better use of the radio resources and reducing interference
- Complex & costly
 - Intensive research related to military, space, etc. activities
 - » Smart antennas, signal-processing antennas, tracking antennas, phased arrays, etc.

Source: adapted from N Gregorieva

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Satellite antennas (TV)



• Not an array!

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Owens Valley Radio Observatory



The Earth's atmosphere is transparent in the narrow visible-light window (4000-7000 angstroms) and the radio band between 1 mm and 10 m.

[Sky & Telescope Feb 1997 p.26]

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The New Mexico Very Large Array



[Sky & Telescope Feb 1997 p. 30]

27 antennas along 3 railroad tracks provide baselines up to 35 km. Radio images are formed by correlating the signals garnered by each antenna.

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2 GHz adaptive antenna



- A set of 48
 2GHz
 antennas
 - Source:Arraycomm

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Phased Arrays

- Array of N antennas in a linear or twodimensional configuration + beam-forming & control device
- The amplitude and phase excitation of each individual antenna controlled electronically ("software-defined")
 - Diode phase shifters
 - Ferrite phase shifters
- Inertia-less beam-forming and scanning (μsec) with fixed physical structure

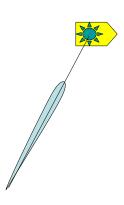
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- Switched beam antennas
 - Based on switching function between separate directive antennas or predefined beams of an array
- Space Division Multiple Access (SDMA) = allocating an angle direction sector to each user
 - In a TDMA system, two users will be allocated to the same time slot and the same carrier frequency
 - They will be differentiated by different direction angles

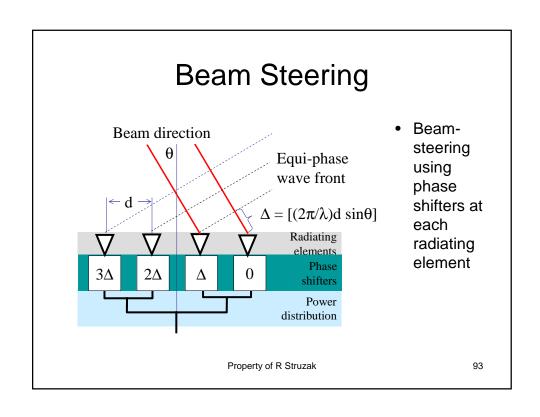
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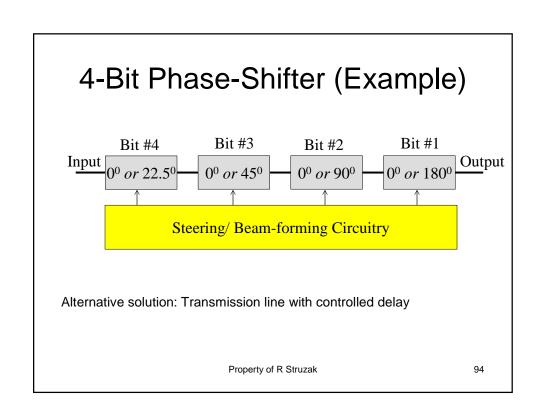
91



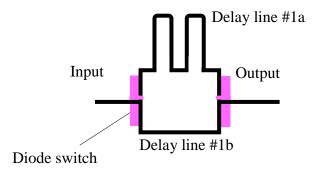
- Dynamically phased array (PA):
 - A generalization of the switched lobe concept
 - The radiation pattern continuously track the designated signal (user)
 - Include a direction of arrival (DoA) tracking algorithm

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Phase bit = delay difference

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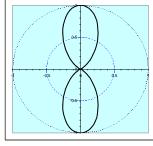
95

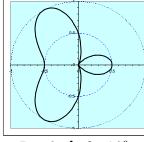
Simulation

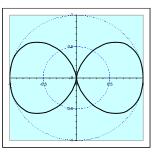
- <u>2 omnidirectional antennas</u> (equal amplitudes)
 - Variables
 - distance increment
 - phase increment
- N omnidirectional antennas
 - Group factor (N omnidirectional antennas uniformly distributed along a straight line, equal amplitudes, equal phase increment)

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2 omnidirectional antennas







$$D = 0.5\lambda$$
, $\theta = 0^0$

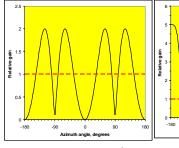
 $D = 0.5\lambda, \theta = 90^{\circ}$

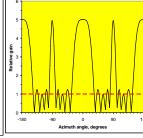
 $D = 0.5\lambda$, $\theta = 180^{\circ}$

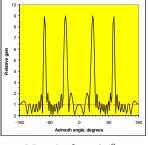
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N omnidirectional antennas







 $N = 2, \theta = 90^{\circ}$

N = 5, $\theta = 180^0$

 $N = 9, \theta = 45^{\circ}$

• Array gain (line, uniform, identical power)

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Antenna Arrays: Benefits

- Possibilities to control electronically
 - Direction of maximum radiation
 - Directions (positions) of nulls
 - Beam-width
 - Directivity
 - Levels of sidelobes

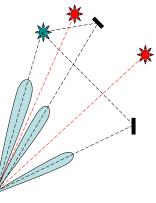
using standard antennas (or antenna collections) independently of their radiation patterns

 Antenna elements can be distributed along straight lines, arcs, squares, circles, etc.

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aa

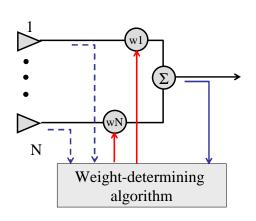
Adaptive ("Intelligent") Antennas



- Array of N antennas in a linear, circular, or planar configuration
- Used for selection signals from desired sources and suppress incident signals from undesired sources
- The antenna pattern track the sources
- It is then adjusted to null out the interferers and to maximize the signal to interference ratio (SIR)
- Able to receive and combine constructively multipath signals

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- The amplitude/ phase excitation of each antenna controlled electronically ("software-defined")
- The weight-determining algorithm uses a-priori and/ or measured information to adapt antenna to changing environment
- The weight and summing circuits can operate at the RF and/ or at an intermediate frequency



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Antenna sitting

- Radio horizon
- Effects of obstacles & structures nearby
- Safety
 - operating procedures
 - Grounding
 - lightning strikes
 - static charges
 - Surge protection
 - · lightning searches for a second path to ground

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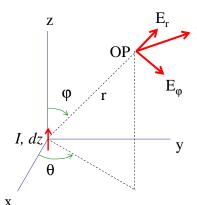
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Maxwell's Equations

- EM field interacting with the matter
 - 2 coupled vectors E and H (6 numbers!), varying with time and space and satisfying the boundary conditions
 (see http://www.amanogawa.com/archive/docs/EM1.pdf;
 http://www.amanogawa.com/archive/docs/EM5.pdf)
- Reciprocity Theorem
 - Antenna characteristics do not depend on the direction of energy flow. The impedance & radiation pattern are the same when the antenna radiates signal and when it receives it.
 - Note: This theorem is valid only for linear passive antennas (i.e. antennas that do not contain nonlinear and unilateral elements, e.g. amplifiers)

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EM Field of Current Element



$$\begin{split} \vec{E} &= \vec{E}_r + \vec{E}_\vartheta + \vec{E}_\varphi \\ \vec{H} &= \vec{H}_r + \vec{H}_\vartheta + \vec{H}_\varphi \end{split}$$

$$\begin{split} \left|E\right| &= \sqrt{\left|E_{r}\right|^{2} + \left|E_{\vartheta}\right|^{2} + \left|E_{\varphi}\right|^{2}} \\ \left|H\right| &= \sqrt{\left|H_{r}\right|^{2} + \left|H_{\vartheta}\right|^{2} + \left|H_{\varphi}\right|^{2}} \end{split}$$

I: current (monochromatic) [A]; *dz*: antenna element (short) [m]

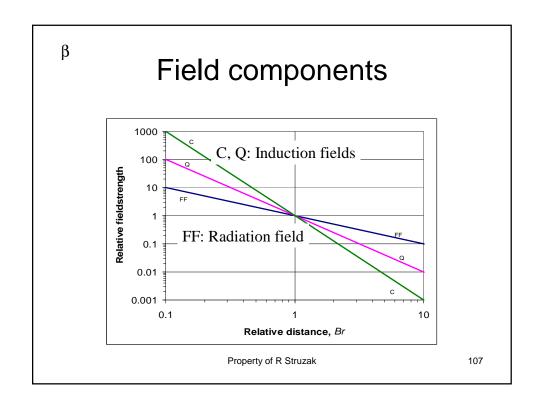
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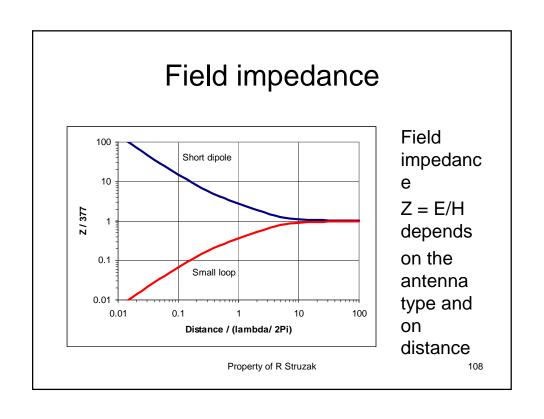
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Short dipole antenna: summary

- $E_{\rm e}$ & $H_{\rm e}$ are maximal in the equatorial plane, zero along the antenna axis
- E_r is maximal along the antenna axis dz, zero in the equatorial plane
- · All show axial symmetry
- All are proportional to the current moment Idz
- Have 3 components that decrease with the distance-to-wavelength ratio as
 - (r/λ)-² & (r/λ)-³: near-field, or induction field. The energy oscillates from entirely electric to entirely magnetic and back, twice per cycle. Modeled as a resonant LC circuit or resonator;
 - $(r/\lambda)^{-1}$: far-field or radiation field
 - These 3 component are all equal at $(r/\lambda) = 1/(2\pi)$

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Far-Field, Near-Field

- Near-field region:
 - Angular distribution of energy depends on distance from the antenna;
 - Reactive field components dominate (L, C)
- Far-field region:
 - Angular distribution of energy is independent on distance;
 - Radiating field component dominates (R)
 - The resultant EM field can locally be treated as uniform (TEM)

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Poynting vector

- The time-rate of EM energy flow per unit area in free space is the *Poynting vector*
 - (see http://www.amanogawa.com/archive/docs/EM8.pdf).
- It is the cross-product (vector product, right-hand screw direction) of the electric field vector (E) and the magnetic field vector (H): P = E x H.
- For the elementary dipole E_θ ⊥ H_θ and only E_θxH_θ carry energy into space with the speed of light.

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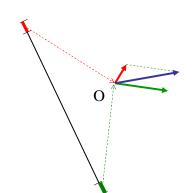
Power Flow

- In free space and at large distances, the radiated energy streams from the antenna in radial lines, i.e. the Poynting vector has only the radial component in spherical coordinates.
- A source that radiates uniformly in all directions is an *isotropic source* (radiator, antenna).
 For such a source the radial component of the Poynting vector is independent of θ and φ.

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Linear Antennas



 Summation of all vector components E (or H) produced by each antenna element

element
$$\vec{E} = \vec{E}_1 + \vec{E}_2 + \vec{E}_3 + \dots$$

$$\vec{H} = \vec{H}_1 + \vec{H}_2 + \vec{H}_3 + \dots$$

- In the far-field region, the vector components are parallel to each other
- Phase difference due to
 - Excitation phase difference
 - Path distance difference
- Method of moments NEC

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Point Source

- For many purposes, it is sufficient to know the direction (angle) variation of the power radiated by antenna at large distances.
- For that purpose, any practical antenna, regardless of its size and complexity, can be represented as a point-source.
- The actual field near the antenna is then disregarded.

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- The EM field at large distances from an antenna can be treated as originated at a point source - fictitious volume-less emitter.
- The EM field in a homogenous unlimited medium at large distances from an antenna can be approximated by an uniform plane TEM wave

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Selected References

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- Stutzman WL, Thiele GA: Antenna Theory and Design JWiley &Sons, 1981
- http://amanogawa.com
- Software

 - http://www.feko.co.za/apl_ant_pla.htm
 Li et al., "Microcomputer Tools for Communication Engineering"
 Pozar D. "Antenna Design Using Personal Computers"
 NEC Archives www.gsl.net/wb6tpu/swindex.html ()

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Java simulations

- Polarization:
 - http://www.amanogawa.com/archive/wavesA.html
- Linear dipole antennas: http://www.amanogawa.com/archive/DipoleAnt/DipoleAnt-2.html
- http://www.amanogawa.com/archive/Antenna1/Antenna1-2.html
- 2 antennas: http://www.amanogawa.com/archive/TwoDipole/Antenna 2-2.html

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Any questions?

Thank you for your attention

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